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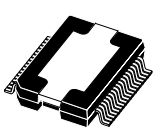
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TDA7575B

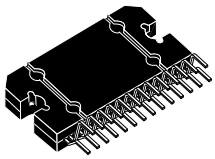
2 x 75W multifunction dual-bridge power amplifier with integrated digital diagnostics

Features

- Multipower bcd technology
- MOSFET output power stage
- DMOS power output
- New high-efficiency (class AB)
- Single-channel 1Ω driving capability
- High output power capability 2x28 W/4 Ω @ 14.4 V, 1 kHz, 10 % THD
- Max. output power 2x75 W/2 Ω, 1x150 W/1 Ω
- Single-channel 1 Ω driving capability
- 84 W undistorted power
- Full I^2C bus driving with 4 address possibilities:
  - Standby
  - Play/mute
  - Gain 12/26 dB
  - Full digital diagnostic (AC and DC loads)
- Possibility to disable the I^2C bus
- Differential inputs
- Full fault protection
- DC offset detection
- Two independent short circuit protections
- Diagnostic on clipping detector with selectable threshold (2 % / 10 %)
- Clipping detector as diagnostic pin when I^2C bus is disabled
- Standby/mute pins
- ESD protection



PowerSO36 (slug up)



Flexiwatt27

Description

The TDA7575B is a new MOSFET dual bridge amplifier specially intended for car radio applications. Thanks to the DMOS output stage the TDA7575B has a very low distortion allowing a clear powerful sound.

Among the features, its superior efficiency performance coming from the internal exclusive structure, makes it the most suitable device to simplify the thermal management in high power sets.The dissipated output power under average listening condition is in fact reduced up to 50% when compared to the level provided by conventional class AB solutions.

This device is equipped with a full diagnostic array that communicates the status of each speaker through the I^2C bus. The TDA7575B has also the possibility of driving loads down to 1Ω paralleling the outputs into a single channel. It is also possible to disable the I^2C and control the TDA7575B by means of the usual standby and mute pins.

Table 1. Device summary

Order code	Package	Packing
TDA7575B	Flexiwatt27	Tube
TDA7575BPD	PowerSO36 (slug up)	Tube
TDA7575BPDTR	PowerSO36 (slug up)	Tape and reel

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1 Block and pins diagrams

Figure 1. Block diagram

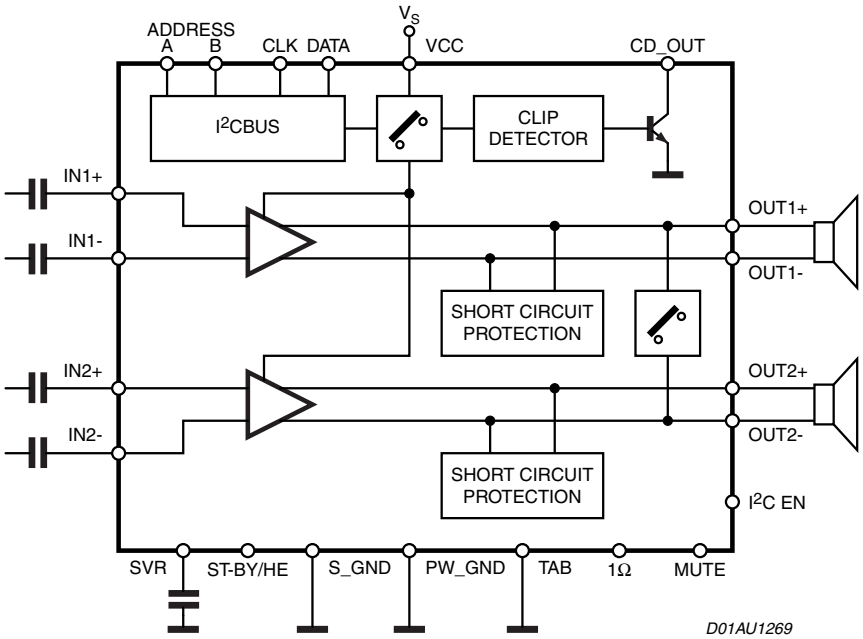
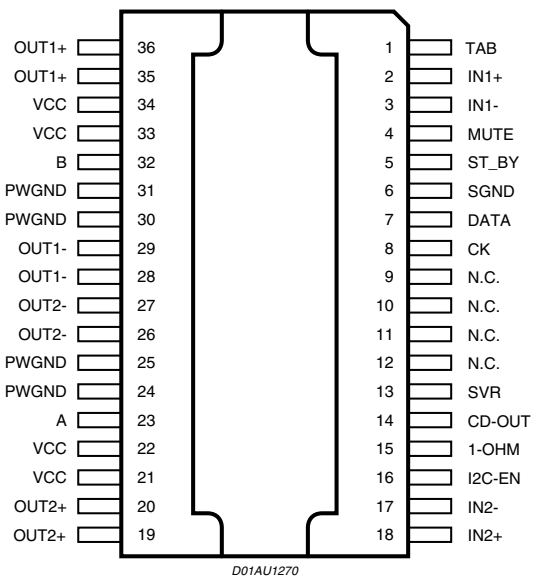
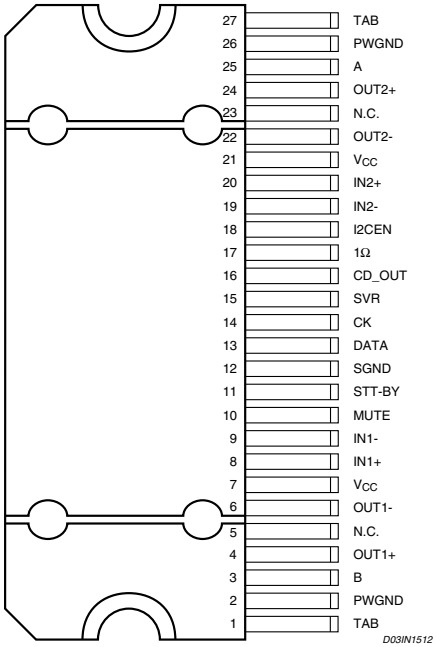


Figure 2. Pins connection diagram (top view)



PowerSO36 (slug up)



Flexiwatt27

2 Electrical specifications

2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{op}	Operating supply voltage	18	V
V_S	DC supply voltage	28	V
V_{peak}	Peak supply voltage (for t = 50 ms)	50	V
V_{CK}	CK pin voltage	6	V
V_{DATA}	Data pin voltage	6	V
I_O	Output peak current (not repetitive t = 100 ms)	8	A
I_O	Output peak current (repetitive f > 10 Hz)	6	A
P_{tot}	Power dissipation T_{case} = 70 °C	86	W
T_{stg}, T_j	Storage and junction temperature	-55 to 150	°C

2.2 Thermal data

Table 3. Thermal data

Symbol	Parameter		PowerSO36	Flexiwatt 27	Unit
$R_{th\ j\text{-}case}$	Thermal resistance junction-to-case	Max	1	1	°C/W

2.3 Electrical characteristics

V_S = 14.4 V; f = 1 kHz; R_L = 4 Ω; T_{amb}= 25 °C unless otherwise specified.

Table 4. Electrical characteristics

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
Power amplifier						
V_S	Supply voltage range	-	8	-	18	V
I_d	Total quiescent drain current	-	50	130	200	mA
P_o	Output power	Max. power(1)	35	40	-	W
		THD = 10 % THD = 1 %; BTL mode	25	28 22	-	W
		R_L = 2 Ω; THD 10 % R_L = 2 Ω; THD 1 % R_L = 2 Ω; Max. power(1)	45 70	50 37 75	-	W

Table 4. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
P_o	Output power	Single channel configuration (1 Ω pin > 2.5 V); R_L = 1 Ω; THD 3 % Max. power(1)	 80 140	 84 150	-	W
THD	Total harmonic distortion	P_o = 1-12 W; STD mode HE mode; P_o = 1-2 W HE mode; P_o = 4-8 W	-	0.03 0.03 0.5	0.1 0.1	%
		P_o = 1-12 W, f = 10 kHz	-	0.15	0.5	%
		R_L = 2; HE mode; P_o = 3 W	-	0.03	0.5	%
		Single channel configuration (1 Ω pin > 2.5 V); R_L = 1; P_o = 4-30 W	-	0.02	0.1	%
C_T	Cross talk	R_g = 600 Ω; P_o = 1 W	60	75	-	dB
R_{IN}	Input impedance	-	60	100	130	kΩ
G_{V1}	Voltage gain 1 (default)	-	25	26	27	dB
ΔG_{V1}	Voltage gain match 1	-	-1	0	1	dB
G_{V2}	Voltage gain 2	-	11	12	13	dB
ΔG_{V2}	Voltage gain match 2	-	-1	0	1	dB
E_{IN1}	Output noise voltage gain 1	R_g = 600 Ω; G_v = 26 dB filter 20 to 22 kHz	-	40	60	µV
E_{IN2}	Output noise voltage gain 2	R_g = 600 Ω; G_v = 12d B filter 20 to 22 kHz	-	15	25	µV
SVR	Supply voltage rejection	f = 100 Hz to 10 kHz; V_r = 1 Vpk; R_g = 600 Ω	50	60	-	dB
BW	Power bandwidth	(-3 dB)	100	-	-	KHz
A_{SB}	Standby attenuation	-	90	100	-	dB
I_{SB}	Standby current consumption	V_{st-by} = 0 V	-	2	10	µA
A_M	Mute attenuation	-	80	90		dB
V_{OS}	Offset voltage	Mute and play	-45	0	45	mV
V_{AM}	Min. supply mute threshold	-	7	7.5	8	V
CMRR	Input CMRR	V_{CM} = 1 Vpk-pk; R_g = 0 Ω	56	60		dB
V_{MC}	Maximum common mode input level	f = 1 kHz	-	-	1	Vrms
SR	Slew rate	-	1.5	4	-	V/µs
ΔV_{OS}	During mute on/off output offset voltage	ITU R-ARM weighted see *Figure 23*	-10	-	+10	mV
	During standby on/off output offset voltage		-10	-	+10	mV

Table 4. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
T_{ON}	Turn on delay	D2 (IB1) 0 to 1	-	15	40	ms
T_{OFF}	Turn off delay	D2 (IB1) 1 to 0	-	15	40	ms
V_{OFF}	Standby pin for standby	-	0	-	1.5	V
V_{SB}	Standby pin for standard bridge	-	3.5	-	5	V
V_{HE}	Standby pin for high-efficiency	-	7	-	18	V
I_O	Standby pin current	$1.5 < V_{st\text{-}by/HE} < 18$ V	7	160	200	µA
	Standby pin current	$V_{st\text{-}by} < 1.5$ V	-10	0	10	µA
V_m	Mute pin voltage for mute mode	-	0	-	1.5	V
V_m	Mute pin voltage for play mode	-	3.5	-	18	V
I_m	Mute pin current (standby)	$V_{mute} = 0$ V, $V_{st\text{-}by} < 1.5$V	-5	0	5	µA
I_m	Mute pin current (operative)	$0 V < V_{mute} < 18$ V, $V_{st\text{-}by} > 3.5$ V	-	65	100	µA
V_{I2C}	I^2C pin voltage for I^2C disabled	-	0	-	1.5	V
V_{I2C}	I^2C pin voltage for I^2C enabled	-	2.5	-	18	V
I^2C	I^2C pin current (standby)	$0V < I^2C$ EN < 18V, $V_{stby} < 1.5V$	-5	0	5	µA
I^2C	I^2C pin current (operative)	I^2C EN <18V, $V_{st\text{-}by}$>3.5V	7	11	15	µA
$V_{1\Omega}$	1 Ω pin voltage for 2ch mode	-	0	-	1.5	V
$V_{1\Omega}$	1 Ω pin voltage for 1 Ω mode	-	2.5	-	18	V
$I_{1\Omega}$	1 Ω pin current (standby)	$0 V < 1 \Omega < 18$ V, $V_{s\text{-}tby} < 1.5$ V	-5	0	5	µA
$I_{1\Omega}$	1 Ω pin current (operative)	$1 \Omega < 18$ V, $V_{st\text{-}by} > 3.5$ V	7	11	15	µA
La	A pin voltage	Low logic level	0	-	1.5	V
Ha		High logic level	2.5	-	18	V
Ia	A pin current (standby)	$0V < A < 18V$, $V_{stby} < 1.5$ V	-5	0	5	µA
Ia	A pin current (operative)	A<18V, $V_{st\text{-}by} > 3.5$V	7	11	15	µA
Lb	B pin voltage	Low logic level	0	-	1.5	V
Hb		High logic level	2.5	-	18	V
Ib	B pin current (standby)	$0 V < B < 18$ V, $V_{s\text{-}tby} < 1.5$ V	-5	0	5	µA
Ib	B pin current (operative)	$B < 18$ V, $V_{st\text{-}by} > 3.5$ V	7	11	15	µA
T_W	Thermal warning	-	-	150	-	°C
T_{PI}	Thermal protection intervention	-	-	170	-	°C
I_{CDH}	Clip pin high leakage current	CD off, $0 V < V_{CD} < 5.5$ V	-15	0	15	µA
I_{CDL}	Clip pin low sink current	CD on; $V_{CD} < 300$ mV	1			mA
CD	Clip detect THD level	D0 (IB1) = 0	0.8	1.3	2.5	%
		D0 (IB1) = 1	5	10	15	%

(*) Standby pin high enables I^2C bus; Standby pin low puts the device in standby condition. (see “prog” for more details)

Table 4. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
Turn-on diagnostics (Power amplifier mode)						
Pgnd	Short to GND det. (below this limit, the output is considered in short circuit to GND)	Power amplifier in standby condition	-	-	1.2	V
Pvs	Short to Vs det. (above this limit, the output is considered in short circuit to VS)	-	V_s -0.9	-	-	V
Pnop	Normal operation thresholds.(within these limits, the output is considered without faults).	-	1.8	-	V_s -1.5	V
Lsc	Shorted load det.	-	-	-	0.5	Ω
Lop	Open load det.	-	130	-	-	Ω
Lnop	Normal load det.	-	1.5	-	70	Ω
Turn-on diagnostics (Line driver mode)						
Pgnd	Short to GND det. (below this limit, the output is considered in short circuit to GND)	Power amplifier in standby	-	-	1.2	V
Pvs	Short to Vs det. (above this limit, the Output is considered in Short Circuit to VS)		V_s -0.9	-	-	V
Pnop	Normal operation thresholds.(within these limits, the output is considered without faults).		1.8	-	V_s -1.5	V
Lsc	Shorted load det.		-	-	1.5	Ω
Lop	Open load det.		400	-	-	Ω
Lnop	Normal load det.		4.5	-	200	Ω
Permanent diagnostics (Power amplifier mode or line driver mode)						
Pgnd	Short to GND det. (below this limit, the output is considered in short circuit to GND)	Power amplifier in Mute or Play condition, one or more short circuits protection activated	-	-	1.2	V
Pvs	Short to Vs det. (above this limit, the Output is considered in Short Circuit to VS)	-	V_s - 0.9	-	-	V
Pnop	Normal operation thresholds.(Within these limits, the Output is considered without faults).	-	1.8	-	V_s -1.5	V
Lsc	Shorted load det.	Pow. amp. mode	-	-	0.5	Ω
		Line driver mode	-	-	1.5	Ω

Table 4. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_O	Offset detection	Power amplifier in play condition AC input signals = 0	±1.5	±2	±2.5	V
I_{NLH}	Normal load current detection	$V_O < (V_S - 5)$pk IB2 (D0) = 0	500	-	-	mA
I_{NLL}	Normal load current detection	$V_O < (V_S - 5)$pk IB2 (D0) = 1	250	-	-	mA
I_{OLH}	Open load current detection	$V_O < (V_S - 5)$pk IB2 (D0) = 0	-	-	250	mA
I_{OLL}	Open load current detection	$V_O < (V_S - 5)$pk IB2 (D0) =1	-	-	125	mA
I^2C bus interface						
f_{SCL}	Clock frequency	-	-	-	400	kHz
V_{IL}	Input low voltage	-	-	-	1.5	V
V_{IH}	Input high voltage	-	2.3	-	-	V

1. Saturated sqare wave output.

3 Electrical characteristics curves

Figure 3. Quiescent drain current vs. supply voltage

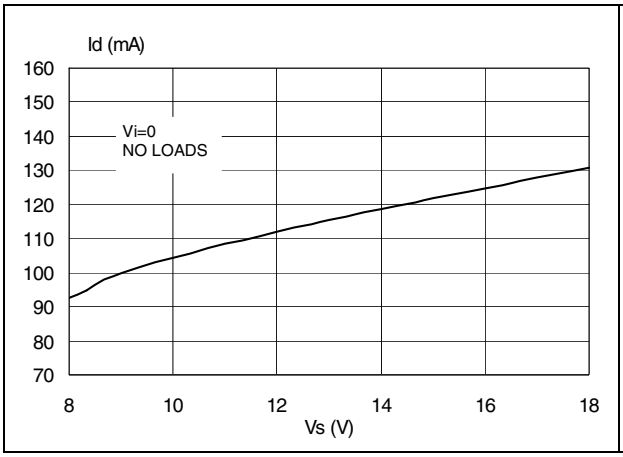


Figure 4. Output power vs. supply voltage

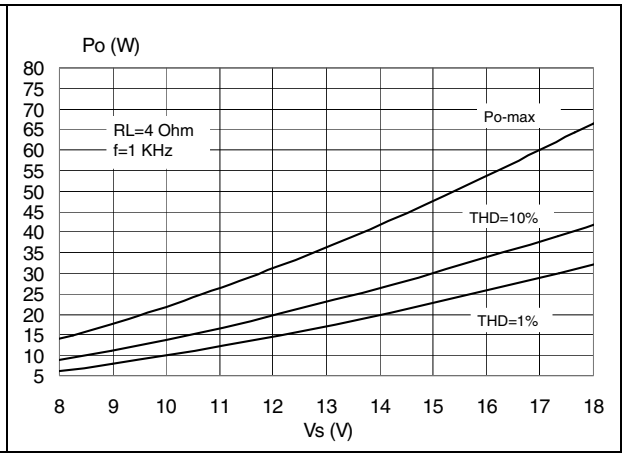


Figure 5. Output power vs. supply voltage

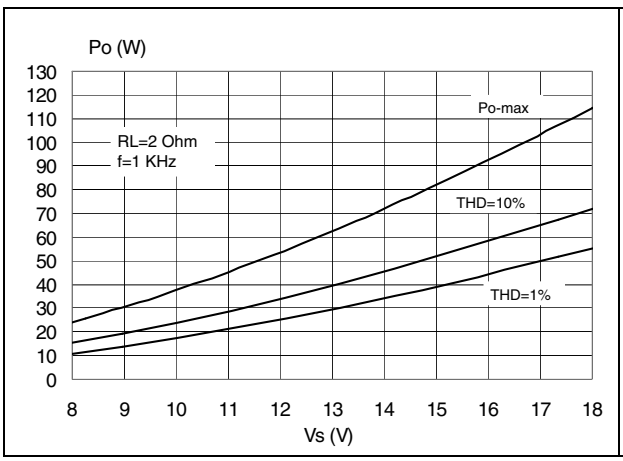


Figure 6. Output power vs. supply voltage

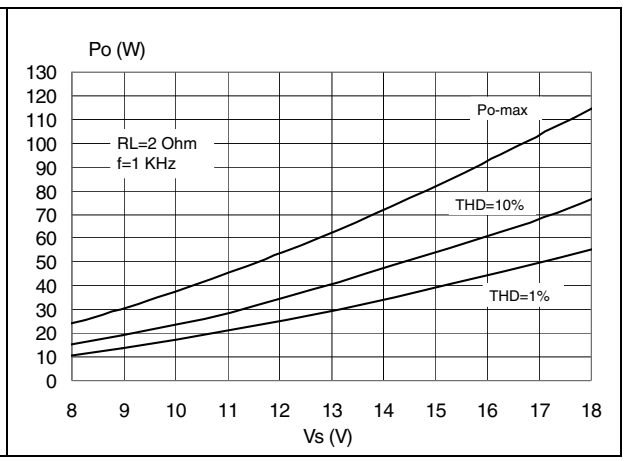


Figure 7. Distortion vs. output power

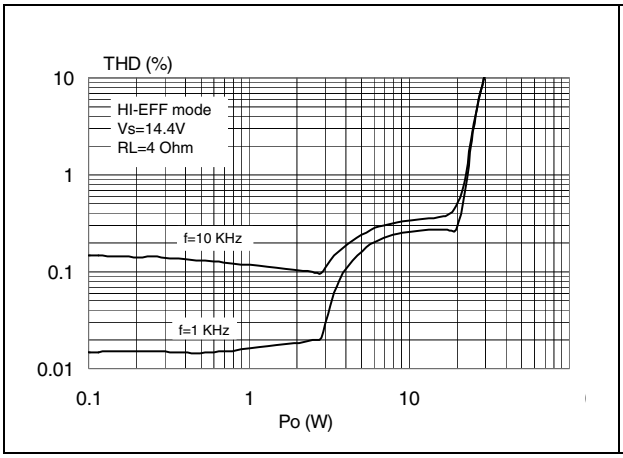


Figure 8. Distortion vs. output power

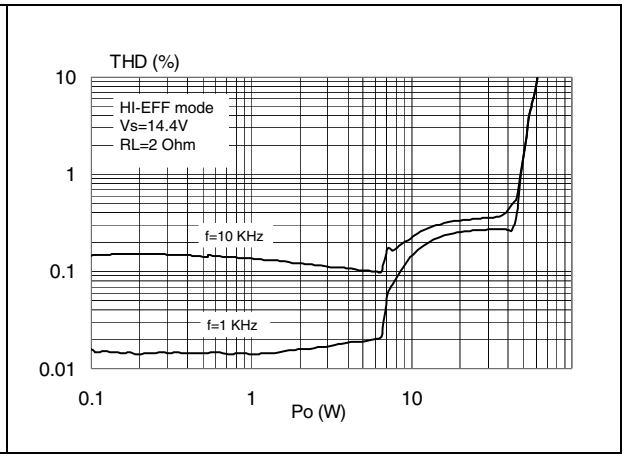


Figure 9. Distortion vs. output power

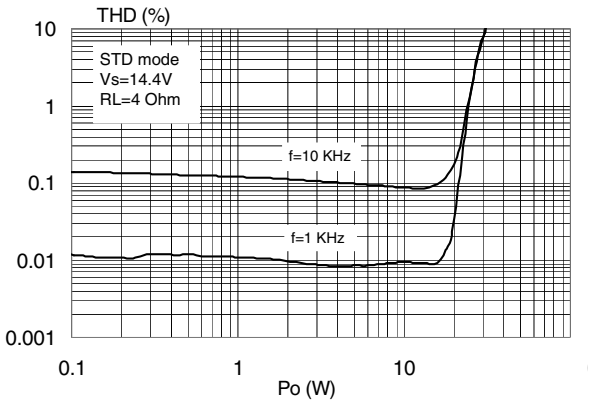


Figure 10. Distortion vs. output power

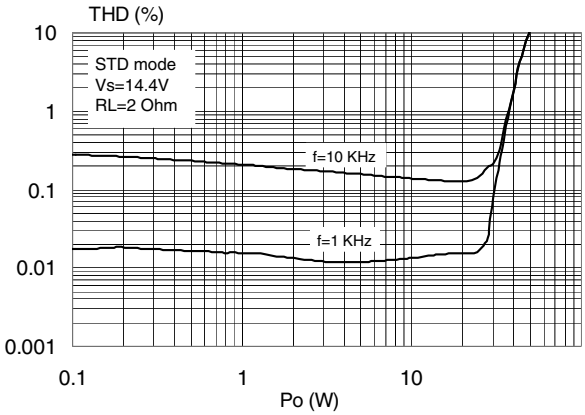


Figure 11. Distortion vs. output power

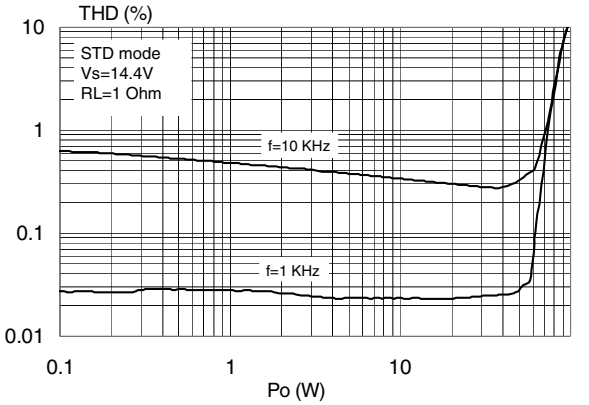


Figure 12. Distortion vs. frequency

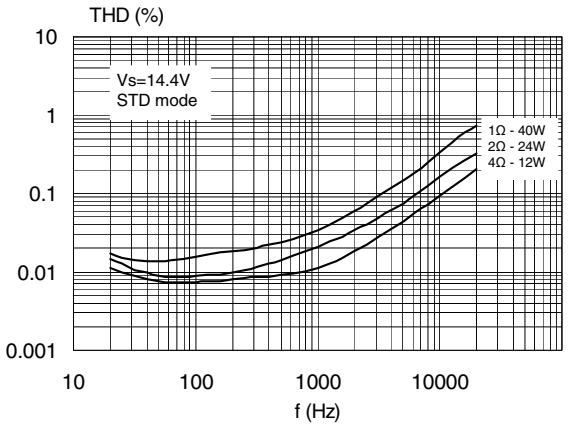


Figure 13. Distortion vs. output voltage (LD mode)

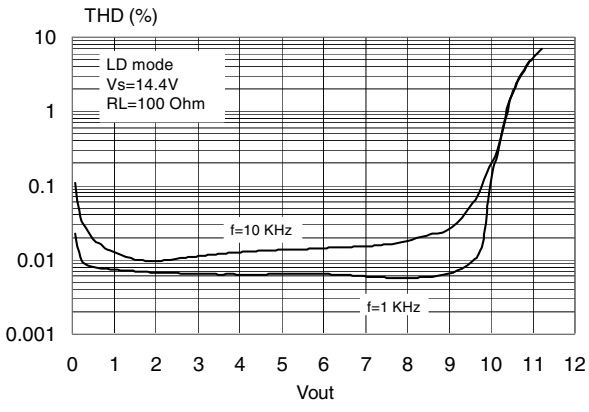


Figure 14. Cross talk vs. frequency

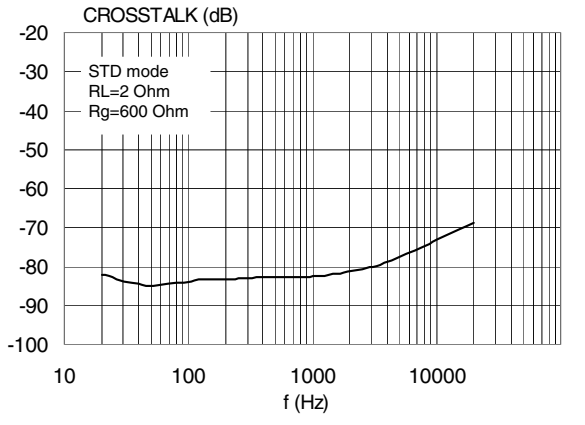


Figure 15. Cross talk vs. frequency (LD mode)

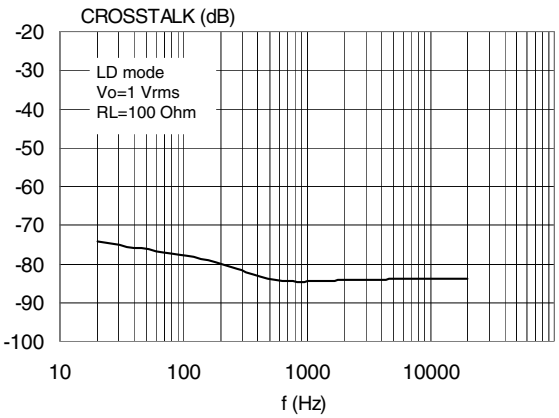


Figure 16. CMRRR vs. frequency

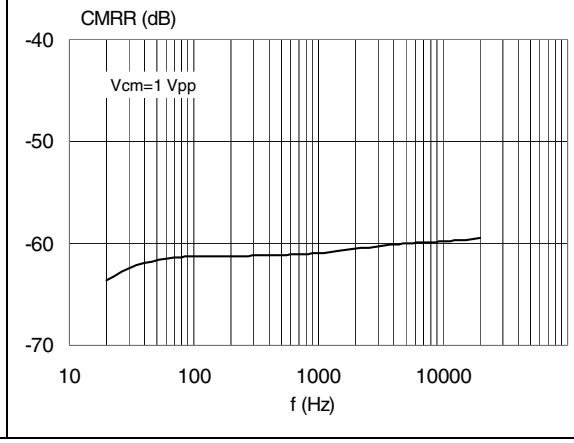


Figure 17. Output attenuation vs. supply voltage (vs. dependent muting)

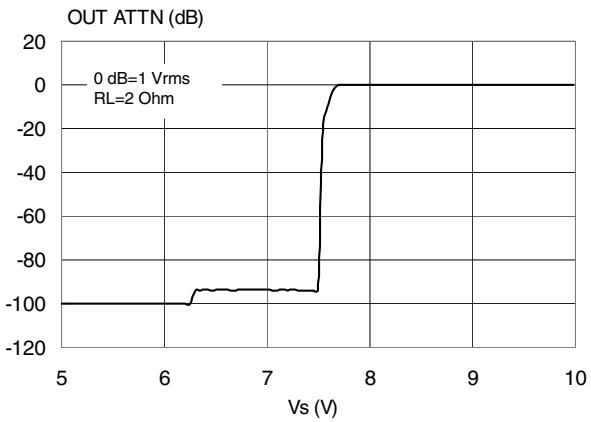


Figure 18. Output attenuation vs. mute pin voltage

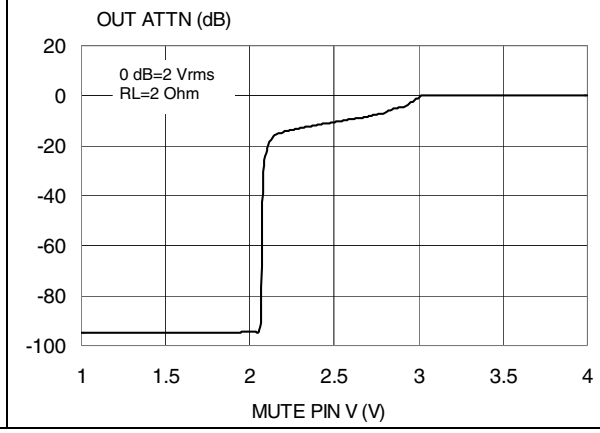


Figure 19. Power dissipation vs. output power (4Ω - SINE)

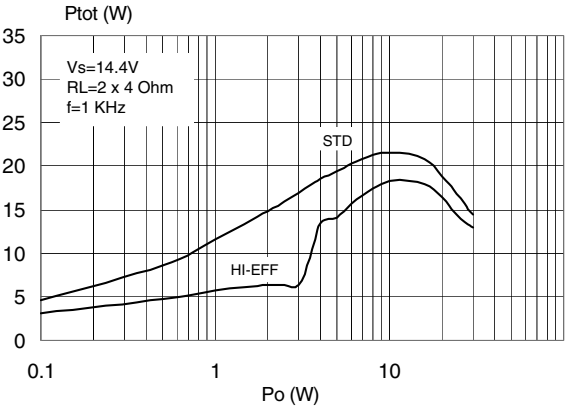


Figure 20. Power dissipation vs. output power (2Ω - SINE)

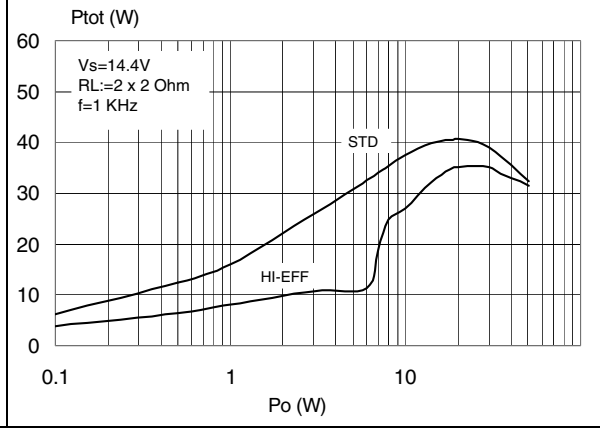


Figure 21. Power dissipation vs. average output power (Audio program simulation, 4Ω)

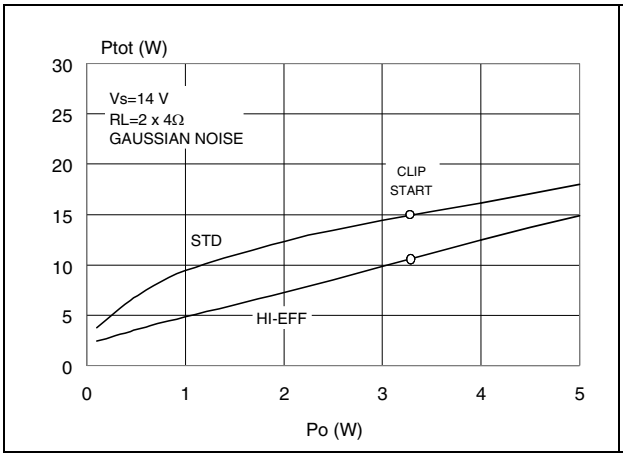


Figure 22. Power dissipation vs. average output power (Audio program simulation, 2Ω)

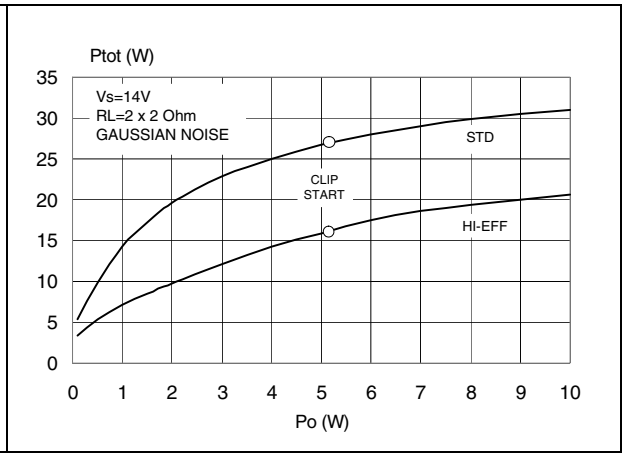
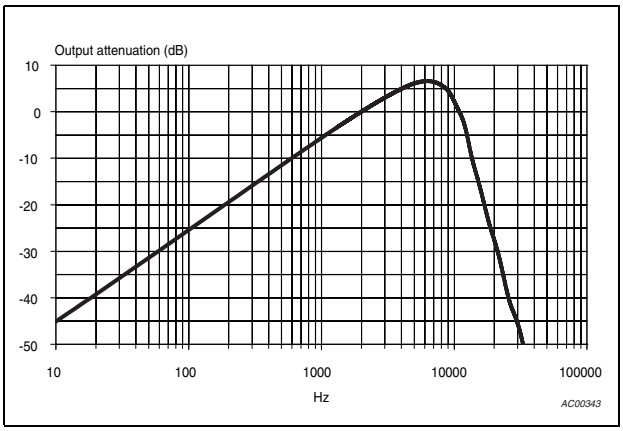


Figure 23. ITU R-ARM frequency response, weighting filter for transient pop



4 Application circuits

Figure 24. Application circuit (TDA7575B)

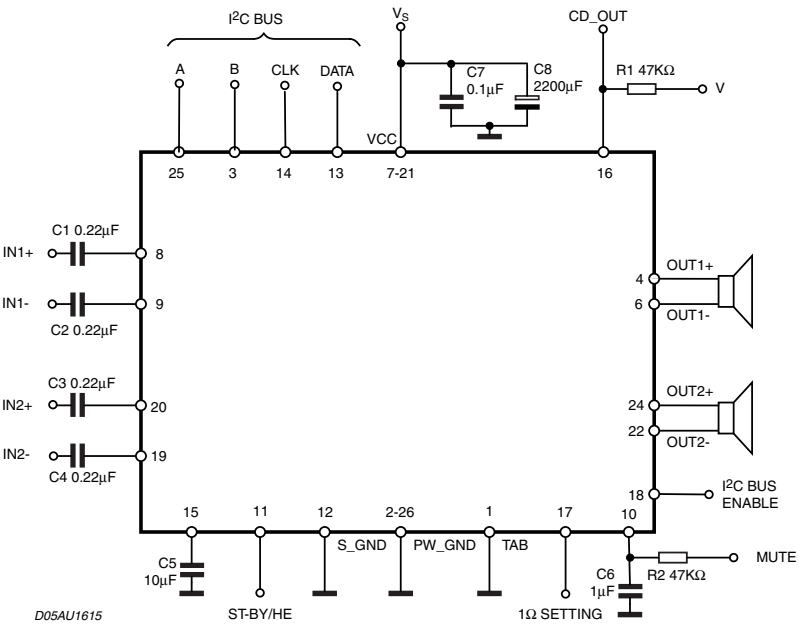
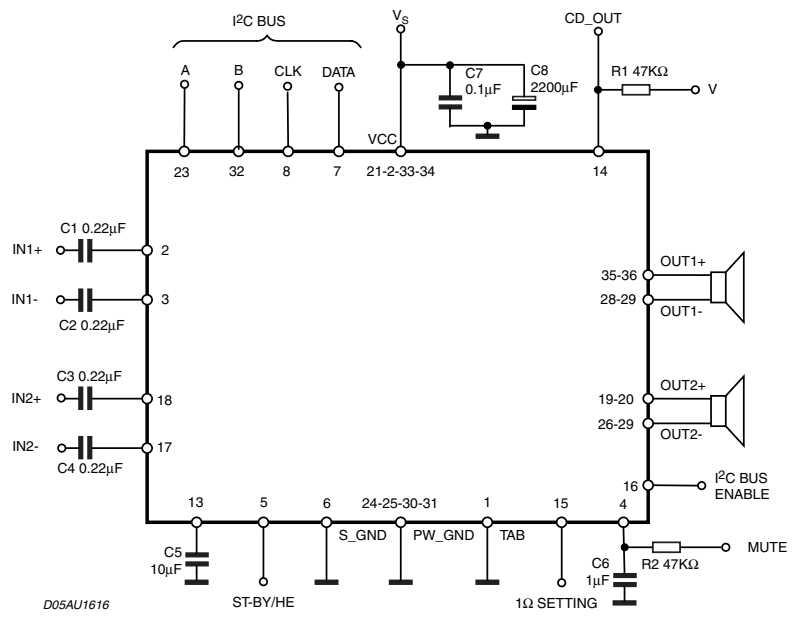


Figure 25. Application circuit (TDA7575BPD)



5 I²C bus interface

Data transmission from microprocessor to the TDA7575B and vice versa takes place through the 2 wires I²C BUS interface, consisting of the two lines SDA and SCL (pull-up resistors to positive supply voltage must be connected).

5.1 Data validity

As shown by *Figure 26*, the data on the SDA line must be stable during the high period of the clock.
The high and low state of the data line can only change when the clock signal on the SCL line is low.

5.2 Start and stop conditions

As shown by *Figure 27* a start condition is a high to low transition of the SDA line while SCL is high.
The stop condltion Is A Low To High Transition of the SDA line while SCL is high.

5.3 Byte format

Every byte transferred to the SDA line must contain 8 bits. Each byte must be followed by an acknowledge bit. The MSB is transferred first.

5.4 Acknowledge

The transmitter(*) puts a resistive HIGH level on the SDA line during the acknowledge clock pulse (see *Figure 28*). The receiver(**) the acknowledges has to pull-down (LOW) the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during this clock pulse.

(*) Transmitter
= master (µP) when it writes an address to the TDA7575B
= slave (TDA7575B) when the µP reads a data byte from TDA7575B

(**) Receiver
= slave (TDA7575B) when the µP writes an address to the TDA7575B
= master (µP) when it reads a data byte from TDA7575B

Figure 26. Data validity on the I²C bus

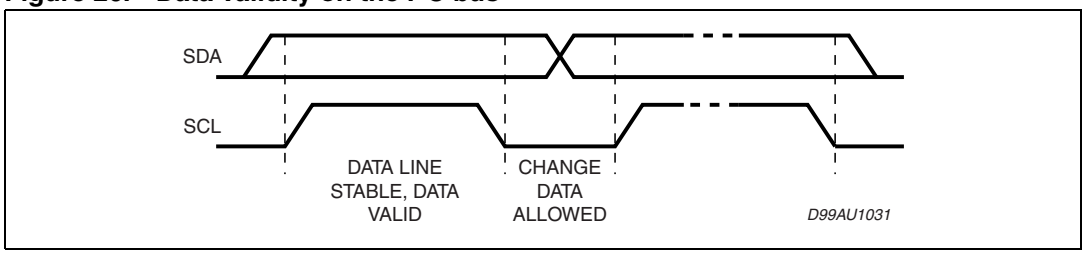


Figure 27. Timing diagram on the I²C bus

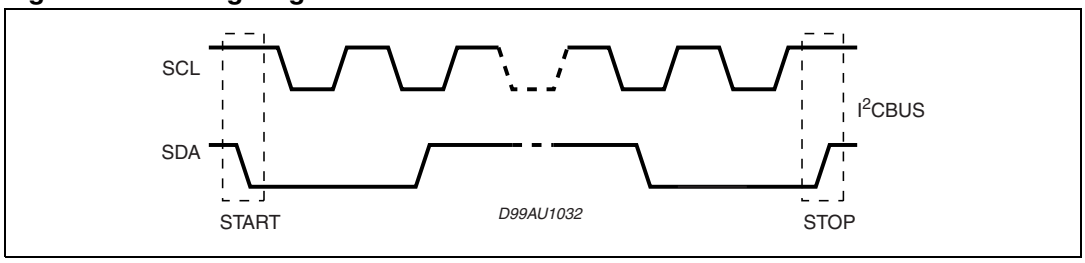
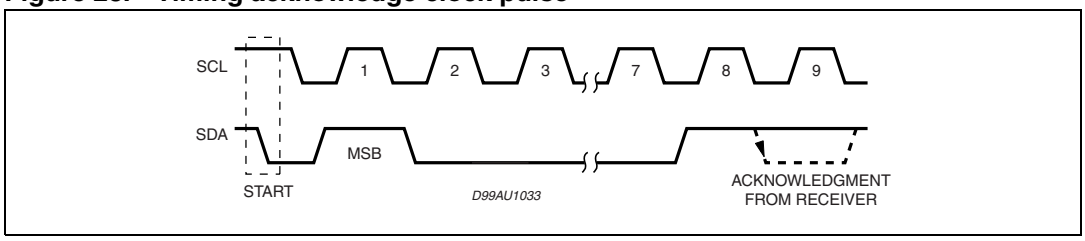


Figure 28. Timing acknowledge clock pulse



5.5 1 Ω capability setting

It is possible to drive 1Ω load paralleling the outputs into a single channel.

In order to implement this feature, outputs are to be connected on the board as follows:

- OUT1+ (pin 35 and pin 36) shorted to OUT2+ (pin 19 and pin 20)
- OUT1- (pin 28 and pin 29) shorted to OUT2- (pin 26 and pin 27).

It is recommended to minimize the impedance on the board between OUT2 and the load in order to minimize THD distortion. It is also recommended to control the maximum mismatch impedance between V_{CC} pins (pin 21/pin 22 respect to pin 33/pin 34) and between PWGND pins (pin 24/pin 25 respect to pin 30/pin 31), mismatch that must not exceed a value of 20 mΩ.

With 1 Ω feature settled the active input is IN2 (pin 17 and pin 18), therefore IN1 pins should be let floating.

It is possible to set the load capability acting on 1 Ω pin as follows:

1 Ω pin (pin 15) < 1.5 V: two channels mode (for a minimum load of 2 Ω)

1 Ω pin (pin 15) > 2.5 V: one channel mode (for 1 Ω load).

It is to remember that 1

Ohm function is a hardware selection.

Therefore it is recommended to leave 1Ω pin floating or shorted to GND to set the two channels mode configuration, or to short 1Ω pin to V_{CC} to set the one channel (1Ω) configuration.

5.6 I^2C abilitation setting

It is possible to disable the I^2C interface by acting on I^2C pin (pin 16) and control the TDA7575B by means of the usual standby and mute pins. In order to activate or deactivate this feature, I^2C pin must be set as follows:

- I^2C pin (pin 16) < 1.5V: I^2C bus interface deactivated
- I^2C pin (pin 16) > 2.5V: I^2C bus interface activated

It is also possible to let I^2C pin floating to deactivate the I^2C bus interface, or to short I^2C pin to V_{CC} to activate it.

In particular:

- I^2C enabled: I^2C pin (pin 16) > 2.5 V
  - STD mode: $V_{st\text{-}by}$ (pin 5) > 3.5 V, IB2(D1)=0
  - HE mode: $V_{st\text{-}by}$ (pin 5) > 3.5 V, IB2(D1)=1
  - Play mode: V_{mute} (pin 4) >3.5 V, IB1 (D2) = 1

The amplifier can always be switched off by putting $V_{st\text{-}by}$ to 0V , but with I^2C enabled it can be turn on only through I^2C (with $V_{st\text{-}by}$ > 3.5 V).

- I^2C disabled: I^2C pin (pin 16) < 1.5 V
  - STD mode: 3.5V < standby (pin 5) < 5
  - HE mode: V_{stby} (pin 5) > 7 V
  - Play mode: V_{mute} (pin 4) > 3.5 V

For both STD and HE mode the play/mute mode can be set acting on V_{mute} pin.

When I^2C bus is disabled, when a fault is detected pin 14 (CD-OUT) is pulled down by the internal logic circuitry. The faults detected are the short circuit to ground, to V_{CC} and across the load (after an aver current detection).

6 Software specifications

All the functions of the TDA7575B are activated by I^2C interface.

The bit 0 of the "Address Byte" defines if the next bytes are write instruction (from µP to TDA7575B) or read instruction (from TDA7575B to µP).

Table 5. Address selection

Bit	Address
A6	1
A5	1
A4	0
A3	1
A2	0
A1	B
A0	A
R/W	X

If R/W = 0, the µP sends 2 "instruction bytes": IB1 and IB2.

Table 6. IB1

Bit	Instruction decoding bit
D7	0
D6	Diagnostic enable (D6 = 1) Diagnostic defeat (D6 = 0)
D5	Offset detection enable (D5 = 1) Offset detection defeat (D5 = 0)
D4	Gain = 26 dB (D4 = 0) Gain = 12 dB (D4 = 1)
D3	0
D2	Mute (D2 = 0) Unmute (D2 = 1)
D1	0
D0	CD 2% (D0 = 0) CD 10% (D0 = 1)

Table 7. IB2

Bit	Instruction decoding bit
D7	0
D6	0
D5	0
D4	Standby on - Amplifier not working - (D4 = 0) Standby off - Amplifier working - (D4 = 1)
D3	Power amplifier mode diagnostic (D3 = 0); Line driver mode diagnostic (D3 = 1)
D2	Current detection diagnostic enabled (D2 = 1) Current detection diagnostic defeat (D2 = 0)
D1	Power amplifier working in standard mode (D1 = 0) Power amplifier working in high efficiency mode (D1 = 1)
D0	Current detection threshold high (D7 =0) Current detection threshold low (D7 =1)

If R/W = 1, the TDA7575B sends 2 "Diagnostics Bytes" to µP: DB1 and DB2.

Table 8. DB1

Bit	Instruction decoding bit	
D7	Thermal warming (if Tchip ≥ 150°C, D7 = 1)	
D6	Diag. cycle not activated or not terminated (D6 = 0) Diag. cycle terminated (D6 = 1)	
D5	Channel 1 current detection IB2 (D0) = 0 Output peak current < 250 mA - Open load (D5 = 1) Output peak current > 500 mA - Normal load (D5 = 0)	Channel LF current detection IB2 (D0) = 1 Output peak current < 125 mA - Open load (D5 = 1) Output peak current > 250 mA - Normal load (D5 = 0)
D4	Channel 1 Turn-on diagnostic (D4 = 0) Permanent diagnostic (D4 = 1)	
D3	Channel 1 Normal load (D3 = 0) Short load (D3 = 1)	
D2	Channel 1 Turn-on diag.: No open load (D2 = 0) Open load detection (D2 = 1) Offset diag.: No output offset (D2 = 0) Output offset detection (D2 = 1)	
D1	Channel 1 No short to V_{cc} (D1 = 0) Short to V_{cc} (D1 = 1)	
D0	Channel 1 No short to GND (D1 = 0) Short to GND (D1 = 1)	

Table 9. DB2

Bit	Instruction decoding bit	
D7	Offset detection not activated (D7 = 0) Offset detection activated (D7 = 1)	
D6	Current sensor not activated (D6 = 0) Current sensor activated (D6 = 1)	
D5	Channel LR Current detection IB2 (D0) = 0 Output peak current < 250 mA - Open load (D5 = 1) Output peak current > 500 mA - Normal load (D5 = 0)	Channel LR Current detection IB2 (D0) = 1 Output peak current < TBD mA - Open load (D5 = 1) Output peak current > TBD mA - Normal load (D5 = 0)
D4	Channel 2 Turn-on diagnostic (D4 = 0) Permanent diagnostic (D4 = 1)	
D3	Channel 2 Normal load (D3 = 0) Short load (D3 = 1)	
D2	Channel 2 Turn-on diag.: No open load (D2 = 0) Open load detection (D2 = 1) Permanent diag.: No output offset (D2 = 0) Output offset detection (D2 = 1)	
D1	Channel 2 No short to V_{cc} (D1 = 0) Short to V_{cc} (D1 = 1)	
D0	Channel 2 No short to GND (D1 = 0) Short to GND (D1 = 1)	

6.1 Examples of bytes sequence

1 - Turn-on diagnostic - Write operation

Start	Address byte with D0 = 0	ACK	IB1 with D6 = 1	ACK	IB2	ACK	STOP

2 - Turn-on diagnostic - Read operation

Start	Address byte with D0 = 1	ACK	DB1	ACK	DB2	ACK	STOP

The delay from 1 to 2 can be selected by software, starting from T.B.D. ms

3a - Turn-on of the power amplifier with mute on, diagnostic defeat.

Start	Address byte with D0 = 0	ACK	IB1	ACK	IB2	ACK	STOP
			X000XXXX		XXX1XX1X		

3b - Turn-off of the power amplifier

Start	Address byte with D0 = 0	ACK	IB1	ACK	IB2	ACK	STOP
			X0XXXXXX		XXX0XXXX		

4 - Offset detection procedure enable

Start	Address byte with D0 = 0	ACK	IB1	ACK	IB2	ACK	STOP
			XX1XX1XX		XXX1XXXX		

5 - Offset detection procedure stop and reading operation (the results are valid only for the offset detection bits (D2 of the bytes DB1, DB2, DB3, DB4).

Start	Address byte with D0 = 1	ACK	DB1	ACK	DB2	ACK	STOP

- The purpose of this test is to check if a D.C. offset (2 V typ.) is present on the outputs, produced by input capacitor with anomalous leakage current or humidity between pins.
- The delay from 4 to 5 can be selected by software, starting from T.B.D. ms

7 Diagnostics functional description

7.1 Turn-on diagnostic

It is activated at the turn-on (stand-by out) under I^2C bus request. Detectable output faults are:

- Short to GND
- Short TO Vs
- Short across the speaker
- Open speaker

To verify if any of the above misconnections are in place, a subsonic (inaudible) current pulse (*Figure 29*) is internally generated, sent through the speaker(s) and sunk back. The Turn On diagnostic status is internally stored until a successive diagnostic pulse is requested (after a I^2C reading).

If the "stand-by out" and "diag. enable" commands are both given through a single programming step, the pulse takes place first (power stage still in stand-by mode, low, outputs = high impedance).

Afterwards, when the Amplifier is biased, the PERMANENT diagnostic takes place. The previous Turn On state is kept until a short appears at the outputs.

Figure 29. Turn-on diagnostic: working principle

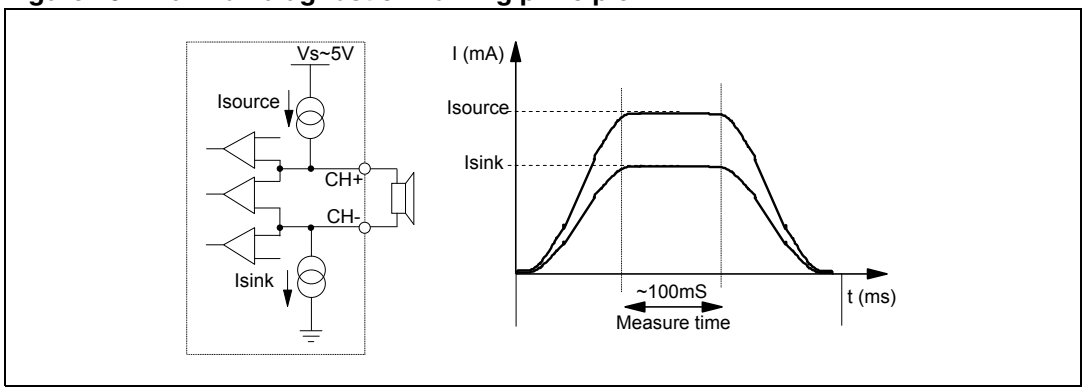


Fig. *Figure 30* and *Figure 31* show SVR and OUTPUT waveforms at the turn-on (stand-by out) with and without Turn-on diagnostic.

Figure 30. SVR and output behavior - case 1: without turn-on diagnostic

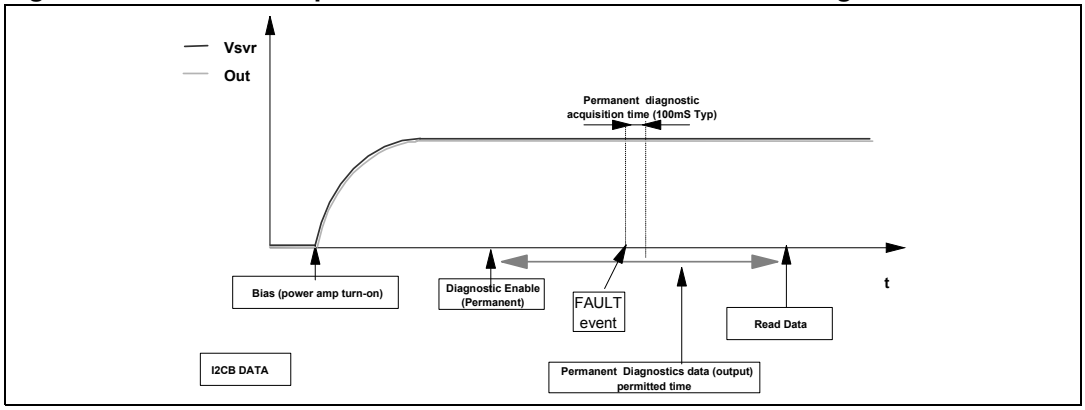
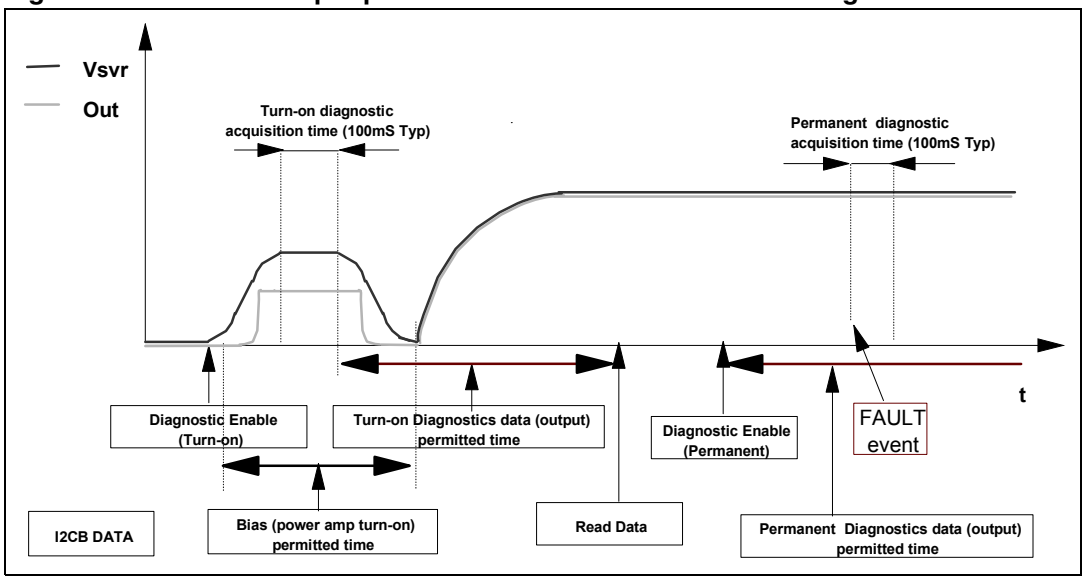
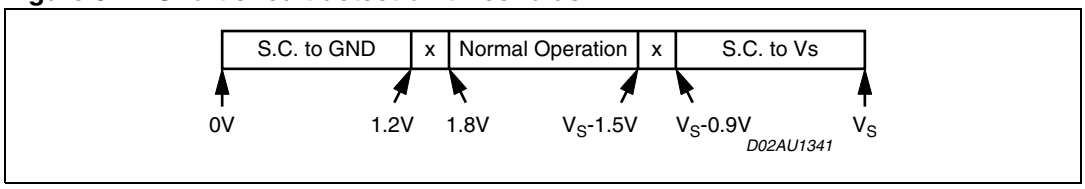


Figure 31. SVR and output pin behavior - case 2: with turn-on diagnostic



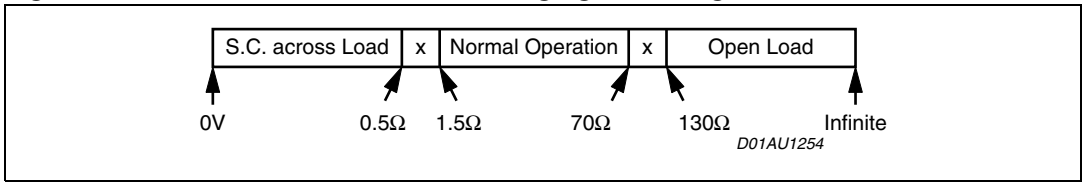
The information related to the outputs status is read and memorized at the end of the current pulse top. The acquisition time is 100 ms (typ.). No audible noise is generated in the process. As for short to GND / Vs the fault-detection thresholds remain unchanged from 26 dB to 12 dB gain setting. They are as follows:

Figure 32. Short circuit detection thresholds



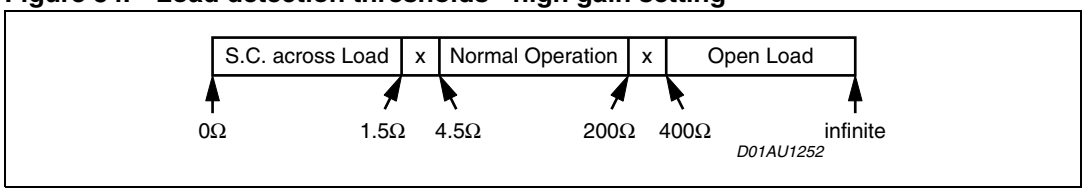
Concerning short across the speaker / open speaker, the threshold varies from 26 dB to 12 dB gain setting, since different loads are expected (either normal speaker's impedance or high impedance). The values in case of 26 dB gain are as follows:

Figure 33. Load detection thresholds - high gain setting



If the line-driver mode (Gv= 12 dB and line driver mode diagnostic = 1) is selected, the same thresholds will change as follows:

Figure 34. Load detection thresholds - high gain setting



7.2 Permanent diagnostics

Detectable conventional faults are:

- Short to GND
- Short to Vs
- Short across the speaker

The following additional features are provided:

- Output offset detection

The TDA7575B has 2 operating statuses:

1. RESTART mode. The diagnostic is not enabled. Each audio channel operates independently from each other. If any of the a.m. faults occurs, only the channel(s) interested is shut down. A check of the output status is made every 1 ms (fig. 30). Restart takes place when the overload is removed.
2. DIAGNOSTIC mode. It is enabled via I^2C bus and self activates if an output overload (such to cause the intervention of the short-circuit protection) occurs to the speakers outputs. Once activated, the diagnostics procedure develops as follows (fig. 31):
   - To avoid momentary re-circulation spikes from giving erroneous diagnostics, a check of the output status is made after 1ms: if normal situation (no overloads) is detected, the diagnostic is not performed and the channel returns back active.
   - Instead, if an overload is detected during the check after 1 ms, then a diagnostic cycle having a duration of about 100 ms is started.
   - After a diagnostic cycle, the audio channel interested by the fault is switched to RESTART mode. The relevant data are stored inside the device and can be read by the microprocessor. When one cycle has terminated, the next one is activated by an I^2C reading. This is to ensure continuous diagnostics throughout the car-radio operating time.
   - To check the status of the device a sampling system is needed. The timing is chosen at microprocessor level (over than half a second is recommended).

Figure 35. Restart timing without diagnostic enable (permanent) each 1ms time, a sampling of the fault is done

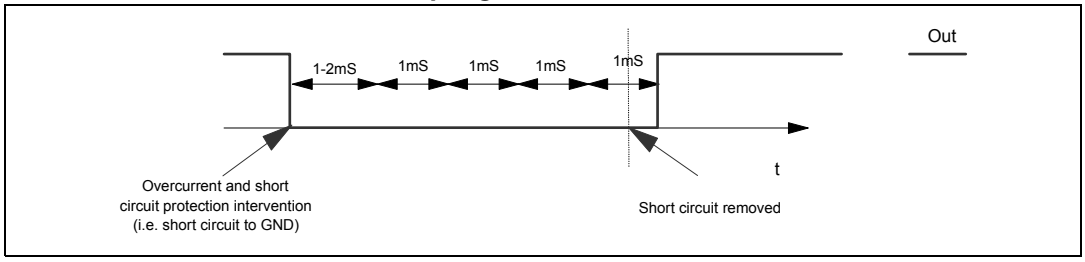
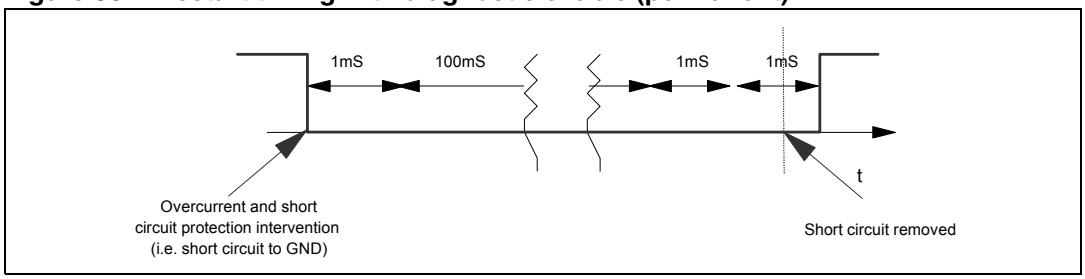


Figure 36. Restart timing with diagnostic enable (permanent)



7.3 Output DC offset detection

Any DC output offset exceeding ± 2 V are signalled out. This inconvenient might occur as a consequence of initially defective or aged and worn-out input capacitors feeding a DC component to the inputs, so putting the speakers at risk of overheating.

This diagnostic has to be performed with low-level output AC signal (or Vin = 0).

The test is run with selectable time duration by microprocessor (from a "start" to a "stop" command):

- Start = Last reading operation or setting IB1 - D5 - (OFFSET enable) to 1
- Stop = Actual reading operation

Excess offset is signalled out if persistent throughout the assigned testing time. This feature is disabled if any overloads leading to activation of the short-circuit protection occurs in the process.

7.4 AC diagnostic

It is targeted at detecting accidental disconnection of tweeters in 2-way speaker and, more in general, presence of capacitively (AC) coupled loads.

This diagnostic is based on the notion that the overall speaker's impedance (woofer + parallel tweeter) will tend to increase towards high frequencies if the tweeter gets disconnected, because the remaining speaker (woofer) would be out of its operating range (high impedance). The diagnostic decision is made according to peak output current thresholds, and it is enabled by setting (IB2-D2) = 1. Two different detection levels are available:

- HIgh current threshold IB2 (D7) = 0
  - Iout > 500 mApk = normal status
  - Iout < 250 mApk = open tweeter
- Low current threshold IB2 (D7) = 1
  - Iout > 250 mApk = normal status
  - Iout < 125 mApk = open tweeter

To correctly implement this feature, it is necessary to briefly provide a signal tone (with the amplifier in "play") whose frequency and magnitude are such to determine an output current higher than 500mApk with IB2(D7)=0 (higher than 250mApk with IB2(D7)=1) in normal conditions and lower than 250 mApk with IB2(D7)=0 (lower than 125 mApk with IB2(D7)=1) should the parallel tweeter be missing.

The test has to last for a minimum number of 3 sine cycles starting from the activation of the AC diagnostic function IB2<D2>) up to the I^2C reading of the results (measuring period). To confirm presence of tweeter, it is necessary to find at least 3 current pulses over the above threholds over all the measuring period, else an "open tweeter" message will be issued.

The frequency / magnitude setting of the test tone depends on the impedance characteristics of each specific speaker being used, with or without the tweeter connected (to be calculated case by case). High-frequency tones (> 10 kHz) or even ultrasonic signals are recommended for their negligible acoustic impact and also to maximize the impedance module's ratio between with tweeter-on and tweeter-off.

Figure 37 shows the load impedance as a function of the peak output voltage and the relevant diagnostic fields.
This feature is disabled if any overloads leading to activation of the short-circuit protection occurs in the process.

Figure 37. Current detection high: load impedance |Z| vs. output peak voltage

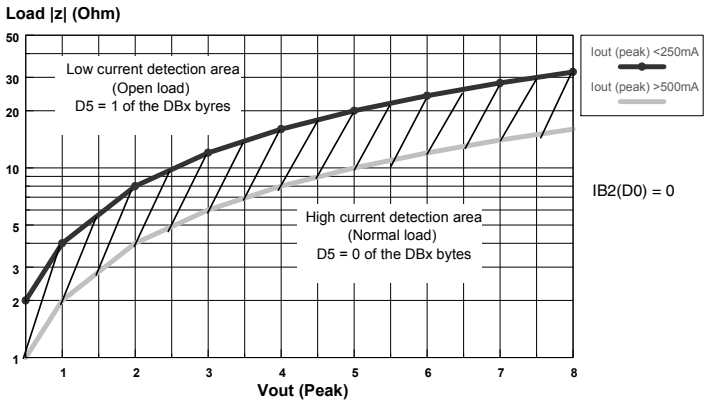
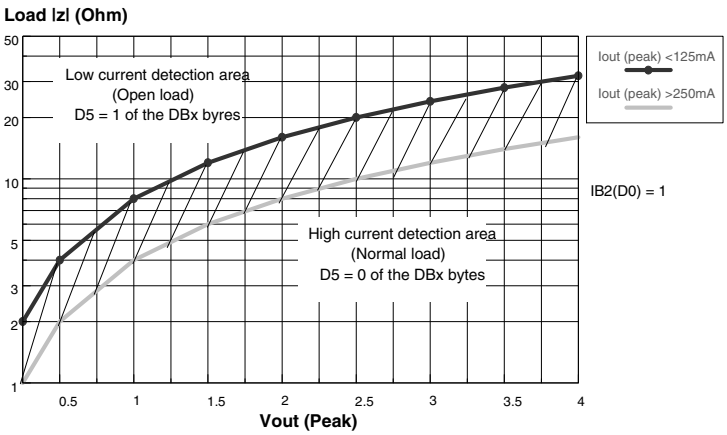


Figure 38. Current detection low: load impedance |Z| vs. output peak voltage



7.5 Multiple faults

When more misconnections are simultaneously in place at the audio outputs, it is guaranteed that at least one of them is initially read out. The others are notified after successive cycles of I^2C reading and faults removal, provided that the diagnostic is enabled. This is true for both kinds of diagnostic (turn-on and permanent).

The table below shows all the couples of double-fault possible. It should be taken into account that a short circuit with the 4 Ω speaker unconnected is considered as double fault.

Table 10. Double fault table for turn-on diagnostic

	S. GND (sc)	S. GND (sk)	S. Vs	S. Across L.	Open L.
S. GND (sc)	S. GND	S. GND	S. Vs + S. GND	S. GND	S. GND
S. GND (sk)	/	S. GND	S. Vs	S. GND	Open L. (*)
S. Vs	/	/	S. Vs	S. Vs	S. Vs
S. Across L.	/	/	/	S. Across L.	N.A.
Open L.	/	/	/	/	Open L. (*)

S. GND (so) / S. GND (sk) in the above table make a distinction according to which of the 2 outputs is shorted to ground (test-current source side= so, test-current sink side = sk). More precisely, in both the channels SO = CH+, and SK = CH-.

In permanent diagnostic the table is the same, with only a difference concerning open load (*), which is not among the recognizable faults. Should an open load be present during the device's normal working, it would be detected at a subsequent turn-on diagnostic cycle (i.e. at the successive car radio turn-on).

7.6 Faults availability

All the results coming from I^2C bus, by read operations, are the consequence of measurements inside a defined period of time. If the fault is stable throughout the whole period, it will be sent out. This is true for DC diagnostic (turn-on and permanent), for offset detector.

To guarantee always resident functions, every kind of diagnostic cycles (turn-on, permanent, offset) will be reactivate after any I^2C reading operation. So, when the micro reads the I^2C, a new cycle will be able to start, but the read data will come from the previous diag. cycle (i.e. The device is in turn-on state, with a short to GND, then the short is removed and micro reads I^2C. The short to GND is still present in bytes, because it is the result of the previous cycle. If another I^2C reading operation occurs, the bytes do not show the short). In general to observe a change in diagnostic bytes, two I^2C reading operations are necessary.

7.7 I^2C programming/reading sequences

A correct turn on/off sequence respectful of the diagnostic timings and producing no audible noises could be as follows (after battery connection):

- Turn-on: (Standby OUT + DIAG enable) --- 500 ms (min) --- muting OUT
- Turn-off: Muting IN --- 20 ms --- (DIAG disable + standby IN)

Car radio installation: DIAG enable (write) --- 20 0ms --- I^2C read (repeat until all faults disappear).

  - Offset test: device in play (no signal)
  - Offset enable - 30 ms - I^2C reading

(repeat I^2C reading until high-offset message disappears).

8 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: *www.st.com*.

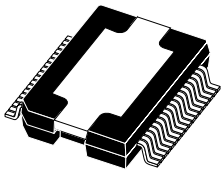
ECOPACK® is an ST trademark.

Figure 39. PowerSO36 (slug up) mechanical data and package dimensions

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	3.270	-	3.410	0.1287	-	0.1343
A2	3.100	-	3.180	0.1220	-	0.1252
A4	0.800	-	1.000	0.0315	-	0.0394
A5	-	0.200	-	-	0.0079	-
a1	0.030	-	-0.040	0.0012	-	-0.0016
b	0.220	-	0.380	0.0087	-	0.0150
c	0.230	-	0.320	0.0091	-	0.0126
D	15.800	-	16.000	0.6220	-	0.6299
D1	9.400	-	9.800	0.3701	-	0.3858
D2	-	1.000	-	-	0.0394	-
E	13.900	-	14.500	0.5472	-	0.5709
E1	10.900	-	11.100	0.4291	-	0.4370
E2	-	-	2.900	-	-	0.1142
E3	5.800	-	6.200	0.2283	-	0.2441
E4	2.900	-	3.200	0.1142	-	0.1260
e	-	0.650	-	-	0.0256	-
e3	-	11.050	-	-	0.4350	-
G	0	-	0.075	0	-	0.0031
H	15.500	-	15.900	0.6102	-	0.6260
h	-	-	1.100	-	-	0.0433
L	0.800	-	1.100	0.0315	-	0.0433
N	-	-	10°	-	-	10°
s	-	-	8°	-	-	8°

(1) "D and E1" do not include mold flash or protusions.
Mold flash or protusions shall not exceed 0.15mm (0.006").
(2) No intrusion allowed inwards the leads.

OUTLINE AND MECHANICAL DATA



PowerSO36 (SLUG UP)

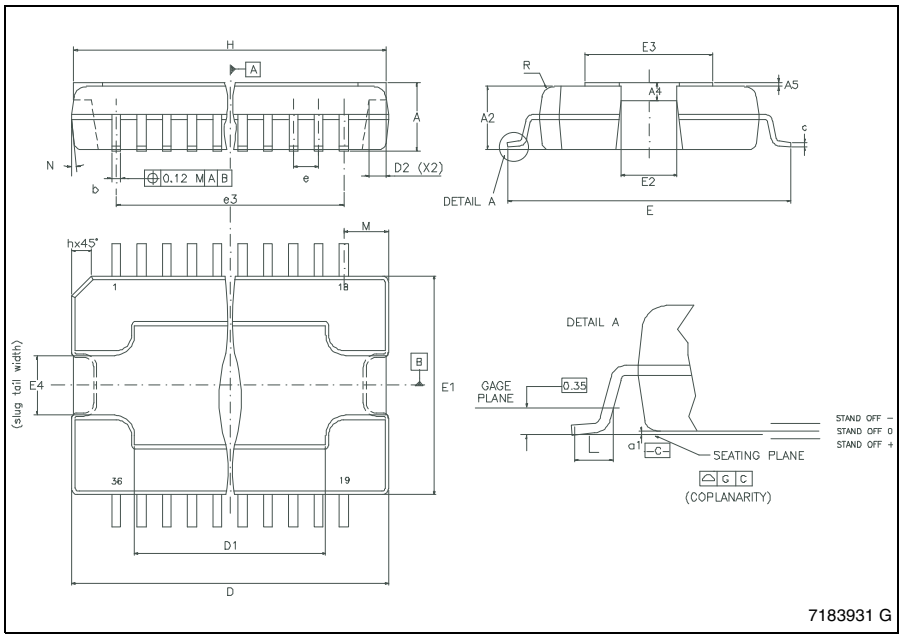
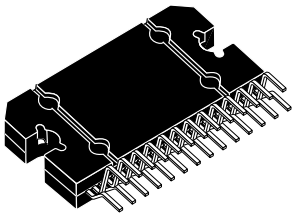


Figure 40. Flexiwatt27 (vertical) mechanical data and package dimensions

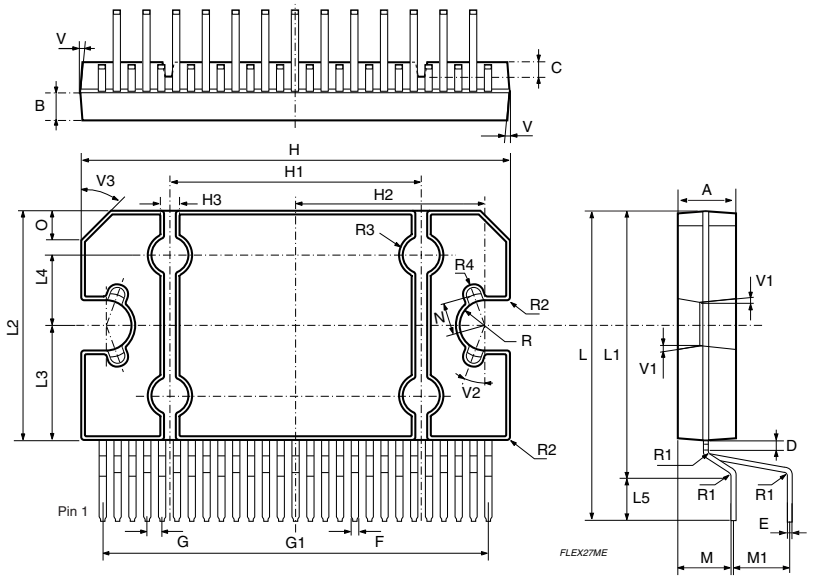
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	4.45	4.50	4.65	0.175	0.177	0.183
B	1.80	1.90	2.00	0.070	0.074	0.079
C		1.40			0.055	
D	0.75	0.90	1.05	0.029	0.035	0.041
E	0.37	0.39	0.42	0.014	0.015	0.016
F (1)			0.57			0.022
G	0.80	1.00	1.20	0.031	0.040	0.047
G1	25.75	26.00	26.25	1.014	1.023	1.033
H (2)	28.90	29.23	29.30	1.139	1.150	1.153
H1		17.00			0.669	
H2		12.80			0.503	
H3		0.80			0.031	
L (2)	22.07	22.47	22.87	0.869	0.884	0.904
L1	18.57	18.97	19.37	0.731	0.747	0.762
L2 (2)	15.50	15.70	15.90	0.610	0.618	0.626
L3	7.70	7.85	7.95	0.303	0.309	0.313
L4		5			0.197	
L5		3.5			0.138	
M	3.70	4.00	4.30	0.145	0.157	0.169
M1	3.60	4.00	4.40	0.142	0.157	0.173
N		2.20			0.086	
O		2			0.079	
R		1.70			0.067	
R1		0.5			0.02	
R2		0.3			0.12	
R3		1.25			0.049	
R4		0.50			0.019	
V	5° (Typ.)					
V1	3° (Typ.)					
V2	20° (Typ.)					
V3	45° (Typ.)					

(1): dam-bar protusion not included
(2): molding protusion included

OUTLINE AND MECHANICAL DATA



Flexiwatt27 (vertical)



9 Revision history

Table 11. Document revision history

Date	Revision	Changes
30-Oct-2007	1	Initial release.
17-Dec-2009	2	Updated *Figure 39: PowerSO36 (slug up) mechanical data and package dimensions on page 29*.
17-Sep-2013	3	Updated Disclaimer.

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